# Device Performance of Graphene Nanoribbon MOSFET and Tunneling FET with Phonon Scattering: A Computation Study

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## 1. Introduction

Research in carbon electronics has been intensified since 2004 and in particular the use of graphene nanoribbon (GNR) in digital devices has been widely explored. While many studies of the GNR metal-oxide-semiconductor field-effect transistors (MOSFET) [1]-[3] and tunneling FET (TFET) [4], [5] has discuss their device performance in the ballistic limits individually, few studies have compared their characteristics with the consideration of phonon scattering [6], [7]. In this study, we present our investigation of the device performance of GNR MOSFET and TFET using a non-equilibrium Green's function (NEGF) quantum simulator with a mode-space Dirac equation model. We observed that in the presence of phonon scattering, both the OFF-state and ON-state currents (IOFF and  $I_{\text{ON}})$  of MOSFET are degraded, and the  $I_{\text{ON}}$  of TFET was enhanced. We also examined the effect of channel length  $(L_{\rm C})$  on the device performances in the presence of phonon scattering and we observed that the I<sub>ON</sub> of MOSFET is more degraded by increasing L<sub>C</sub> than that of TFET.

## 2. Methodology

A schematic of the device simulated in this study is shown in Fig. 1(a). A GNR of width 1.0 nm is chosen, with an energy bandgap of 0.87 eV. The source is n-type doped for MOSFET and p-type doped for TFET and the drain is n-type doped for both devices. The doping concentration is  $4.47 \times 10^{13}$  cm<sup>-2</sup> for both n- and p-type doping. The device has a double metal gate structure with silicon dioxide as the insulating material. The retarded Green's function  $G^r$  in the NEGF formulism is obtained from [8]

$$G^{r} = \left[ EI + i\eta - H_0 - U - \Sigma_S - \Sigma_D - \Sigma_{ph} \right]^{-1}.$$
 (1)

The device Hamiltonian  $H_0$  is described with the mode-space Dirac equation [4] and the  $\Sigma_S$  and  $\Sigma_D$  are the self-energies of the source and drain. The phonon scattering is incorporated in the simulator via  $\Sigma_{ph}$  which is obtained by the sum of in- and out-scattering self-energies calculated based on the self-consistent Born approximation [8], [9]

$$\Sigma_{ph}^{in/out}(E) = \int \left\{ D^{ab}(\hbar\omega) \left[ N(\hbar\omega) + 1 \right] G^{n/p}(E \pm \hbar\omega) \right. \\ \left. + D^{em}(\hbar\omega) N(\hbar\omega) G^{n/p}(E \mp \hbar\omega) \right\} d\omega.$$
(2)



Fig. 1 (a) Schematic of the device simulated. The oxide thickness  $(t_{ox})$  is 1 nm and the gate thickness  $(t_{gate})$  is 3 nm. The band diagram along the device transport direction at  $V_{DS} = 0.4$  V and  $V_{GS}-V_{fb} = 0$  V are shown in (b) and (c) for GNR MOSFET and TFET, respectively. The  $V_{fb}$  is the flat band potential due to the metal gate work function and is set to be 0.2 V here. The transfer characteristics of GNR MOSFETS and TFET for ballistic (BALL) and phonon scattering (APOP) regimes are shown in (d) and (e), respectively. Insets show the linear plots at high gate biases.

The electron-phonon coupling constants  $D^{ab} = D^{em} = D$ is obtained following Ref. 6 and for acoustic phonon (AP),  $D_{AP} = 3.1 \times 10^{-3} \text{ eV}^2$  and for optical phonon (OP), assuming only phonon energy  $\hbar \omega = 0.19 \text{ eV}$  is significant,  $D_{OP} =$  $1.30 \times 10^{-2} \text{ eV}^2$ . *N* is the phonon occupation number from the Bose-Einstein distribution at room temperature and  $G^{n/p}$  is the electron/hole correlation function [8]. The potential *U* is calculated self-consistently by a 2D Poisson solver [5].

### 3. Results and discussion

The transfer characteristics of the GNR MOSFET and TFET are shown in Fig. 1(a) and 1(b) respectively with different  $V_{DS}$ , with  $L_C = 16$  nm. For MOSFET, the phonon scattering increased the  $I_{OFF}$  and decreased the  $I_{ON}$ , with 76%-82% ballisticity. However, the  $I_{OFF}$  of TFET was increased by 4 orders while the  $I_{ON}$  for  $V_{DS} = 0.4$  V was nearly doubled. In order to understand the different effect of phonon scattering on the devices, the IV and the current flux plots of the devices under  $V_{DS} = 0.4$  V are plotted in Fig. 2 and 3 for GNR MOSFET and TFET, respectively.



Fig. 2 (a) The IV of MOSFET at  $V_{DS} = 0.4$  V at low gate biases. (b) and (c) show the current flux plots at  $V_{GS}$ - $V_{fb} = 0$  V. (d) The IV of MOSFET at high gate biases and (e) and (f) show the current flux plots at  $V_{GS}$ - $V_{fb} = 0.8$  V. The arrows show phonon emission (downwards) and absorption (upwards).



Fig. 3 (a) The IV of TFET at  $V_{DS} = 0.4$  V at low gate biases. (b) and (c) show the current flux plots at  $V_{GS}$ - $V_{fb} = 0$  V. (d) The IV of TFET at high gate biases and (e) and (f) show the current flux plots at  $V_{GS}$ - $V_{fb} = 0.8$  V. In (f), the dash arrow indicates a small current flowing from drain to source due to the huge accumulation of charges, reducing the total current. This 'back tunneling' current is reduced for larger channel length.

For GNR MOSFET, AP scattering did not affect the currents at low gate biases [Fig. 2(a)] due to the high barrier where only a very small current was present, and hence back scattering was not apparent. However, OP scattering

increased the amount of carrier at the source by phonon absorption and a larger  $I_{OFF}$  was observed [Fig. 2(c)]. At higher gate biases, the back scattering reduced the current significantly [Fig. 2(d)]. Conversely, OP scattering reduced the back scattering by bringing down the carriers' energy via phonon emission in the channel [Fig. 2(f)] and the current was restored. However, the  $I_{ON}$  was still lower than the ballistic values due to a finite amount of back scattering.

For GNR TFET, AP scattering did not affect the current at all gate biases [Fig. 3(a) & 3(d)] due to the higher carrier concentration at the source than the channel, which prohibited back scattering. On the other hand, OP scattering enhanced band-to-band tunneling (BTBT) [Fig. 3(c) & 3(f)] which increased the current significantly. In the presence of OP scattering, BTBT between source and channel became the dominant mechanism even for  $I_{OFF}$  [Fig. 3(c)].



Fig. 4 The channel length dependence of (a) I<sub>OFF</sub> and (b) I<sub>ON</sub>.

Lastly, the effect of L<sub>C</sub> for GNR MOSFET and TFET was investigated and the  $I_{\text{OFF}}$  (V\_{\text{GS}}\text{-}V\_{\text{fb}} = 0 V) and  $I_{\text{ON}}$  $(V_{GS}-V_{fb} = 0.8 \text{ V})$  as a function of LC is plotted in Fig. 4. While the IOFF of both devices increased with phonon scattering, the I<sub>OFF</sub> of TFET was still lower than MOSFET. It was noted that the ballistic I<sub>OFF</sub> for TFET was decreasing with increasing L<sub>C</sub>, as the dominant mechanism was BTBT between source and drain. With phonon scattering, BTBT occurs between the source and channel and hence IOFF remained invariant. The ION of GNR MOSFET decreased with increasing L<sub>C</sub> due to the increased back scattering and since AP scattering had minimal effect on GNR TFET, the I<sub>ON</sub> was in general not affected by L<sub>C</sub>. However, due to the accumulation of charges at the drain, a minute amount of carrier tunneling from the drain to the source occurs [Fig. 3(f)]. As L<sub>C</sub> increased, this 'back tunneling' was reduced and hence the I<sub>ON</sub> increased initially.

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