

# CMOS Readout Circuit with an On-chip Offset Voltage for Temperature Compensation of pH-ISFET Sensor

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## Abstract

This paper presents a thermally stable CMOS pulse-width-modulation (PWM) readout circuit in which an offset voltage,  $V_{os}$ , is used for temperature compensation of the ion-sensitive field effect transistor (ISFET) of a pH sensor. The effective gate voltage,  $V_{G,eff}$ , of the ISFET in a phosphate-buffered saline (PBS) solution usually has a negative temperature coefficient. The  $V_{os}$  of the readout circuit is used to partially offset the input voltage,  $V_{G,eff}$ . The  $V_{os}$  can be chosen to be complementary to absolute temperature (CTAT) and hence the temperature drift of the  $V_{G,eff}$  of the ISFET in a solution can be offset. The  $V_{os}$  comes from a multiple-positive-input operational amplifier in order that external or on-chip CTAT offset signals can be chosen. The variation quantity of the output pulse width with temperature is about  $9.5 \mu\text{s}/^\circ\text{C}$  and less than  $0.5 \mu\text{s}/^\circ\text{C}$  by using a constant and a CTAT offset voltage, respectively, over a temperature range from 10 to  $50^\circ\text{C}$ .

## 1. Introduction

An extended-gate ion-sensitive field-effect transistor (ISFET) is a metal-oxide-semiconductor FET (MOSFET) with an extended top-metal gate electrode, on which a sensing film is coated. Under a bias voltage  $V_{ref}$  from a reference electrode such as Ag/AgCl, the  $\text{H}^+$  ion concentration causes an corresponding interface potential on the sensing film that is measured as a change in the threshold voltage or a change in the effective floating-gate voltage,  $V_{G,eff}$  [1, 2].

Several studies have demonstrated that ISFETs have large thermal instability due to characteristics of the sensing films [2-3], which results in inaccurate ion measurements. The temperature characteristics of ISFET sensors complexly depend on the reference electrode, electrolyte-insulator potential and ISFET-based MOS transistor. In this work, a CMOS pulse-width-modulation (PWM) readout circuit with an on-chip temperature-dependent offset voltage,  $V_{os}$ , is used for the temperature compensation of the ISFET of a pH sensor.

## 2. Sensing Readout Circuit Design

Fig.1 shows the circuit schematic of the PWM readout circuit with a  $V_{os}$  for temperature compensation of pH-ISFET [1]. The readout circuit consists of two voltage-to-current (V-I) converter, a comparator module, a charging and discharging module (CDM), an output module, and several current

mirrors. The low-voltage cascoded current mirrors are used to mirror the charge, discharge, and offset currents to the CDM in order to charge or discharge the capacitor  $C_p$ . The transistor at the positive input terminal is an extended-gate ISFET. The native  $\text{Al}_2\text{O}_3$  on the top metal is used as the sensing membrane. Its  $V_{G,eff}$  acts as  $V_{sen}$ , and is converted into a current  $I_{sen}$ , i.e.,  $V_{sen}/R_1$ , by one V-I converter. The current is used to generate a single pulse. Another V-I converter uses a multiple-input operational amplifier (OPA) with two positive inputs  $V_{dsg}$  and  $V_{os}$  to generate a discharge current  $I_{dag}$  or an offset current  $I_{os}$ , respectively. The  $I_{os}$  is used to offset the drift of  $V_{G,eff}$  with temperature. By using the multi-input OPA, the circuits related to offset and discharge functions are combined. To facilitate the usage of various  $V_{os}$  voltage sources, another multiple-input OPA, which inputs come from an external voltage bias or outputs of an on-chip temperature sensors with multiple output voltages [4], is connected to the  $V_{os}$  terminal of the PWM readout circuit. The output of the OPA can be selected from these inputs by two selection signals. The temperature sensor mainly consists of three cascoded diode-connected sub-threshold NMOSFETs which  $V_{GS}$  decreases with increasing temperature under a bias current. Therefore, temperature characteristics of its drain voltages exhibit negative temperature coefficients (TCs) [4].

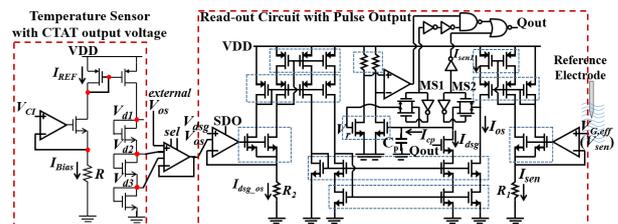


Fig. 1 Circuit schematic of the thermally stable CMOS PWM readout circuit with an on-chip offset voltage.

With two external signals RESET and SW, a specific control signal generator produces two control signals MS1 and MS2 to control transmission-gate (TG) switches of charging and discharging the capacitor  $C_p$  [1]. Initially, MS1 and MS2 are at the "0" and "1" levels, respectively, and the  $C_p$  is charged to have a voltage drop of  $V_r$ . When the signals MS1 and MS2 are at the "1" and "0" levels, respectively, the  $C_p$  is charged by the current  $I_{cp} = I_{sen1} - I_{os}$  for a duration of one half-period of the SW. When the MS2 goes from low level to high level, the Qout also goes from low level to high level and

hence the  $C_p$  is discharged by the current  $I_{dsg}$  until its voltage drop  $V_{Cp}$  goes below the referred  $V_r$ . When the discharging of the  $C_p$  stops, the  $Q_{out}$  goes from high level to low level immediately. By this procedure, a single pulse, which pulse width is linearly proportional to the  $V_{sen}$ , is generated at the  $Q_{out}$ . The MS2 is also used to select  $V_{dsg}$  and  $V_{os}$  as the input of the multiple-input OPA. The circuit principle can be comprehended by referring to refs. [1]. The pulse width  $T_{pw}$  can be derived as follows.

$$(I_{sen1} - I_{os})\left(\frac{1}{2}T_{sw}\right) = I_{dsg}T_{pw} \quad (1)$$

$$T_{pw} = \frac{(V_{sen}/4R_1) - (V_{os}/4R_2)}{(V_{dsg}/16R_3)}\left(\frac{1}{2}T_{sw}\right) = \frac{2(V_{sen} - V_{os})}{V_{dsg}}T_{sw} \quad (2)$$

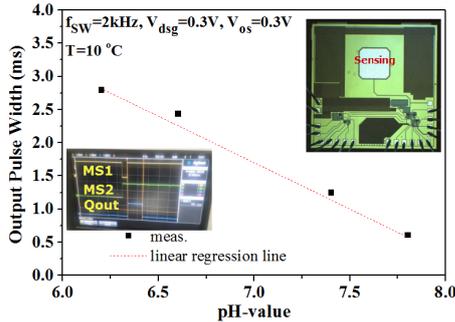


Fig. 2 Measurement result of the output pulse width related to the pH value of the analyzed solution.

Fig. 2 shows the measurement result of the output pulse width related to the pH value of the analyzed solution which is a phosphate-buffered saline (PBS) solution under a bias voltage  $V_{ref}$  of 1.5V, a  $V_{dsg}$  of 0.3V, and  $V_{os}$  of 0.3 V. The SW is a continuous pulse signal of 2 kHz. The inserted photographs are the measured waveform and the chip photograph of the readout circuit chip with a sensing area of  $500 \times 500 \mu m^2$ . The sensitivity is about  $-1.39 \text{ ms/pH}$  with linearity of 99.36%.

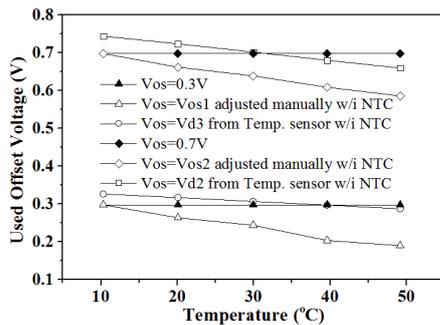


Fig. 3 Temperature characteristics of the used offset voltages for the PWM readout circuit.

Fig. 3 shows the temperature characteristics of the used offset voltages for the PWM readout circuit. These signals are the positive inputs of the multiple-input OPA and can be chosen as the output and then works as the  $V_{os}$  of the PWM readout circuit. 0.3V and 0.7V constant voltages, manually adjusted voltages with a specific negative TC, and the voltages coming from the output voltages of the on-chip temperature sensor are used as the  $V_{os}$ .

Fig. 4 shows the variation of the output  $T_{pw}$  with tempera-

ture under various offset voltages as shown in Fig. 3. The analyzed solution has a pH value of 6.6. With the constant offset voltages of 0.3 V or 0.7 V, the  $T_{pw}$  has a temperature coefficient of about  $-9.5 \mu s/^\circ C$ . By using manually-setting CTAT  $V_{os}$  and on-chip CTAT  $V_{os}$  from  $V_{d2}$  with temperature coefficients of about  $-2.7 \text{ mV}/^\circ C$  and  $-2.1 \text{ mV}/^\circ C$ , respectively, the variations of the output pulse widths are about  $0.15 \mu s/^\circ C$  and  $-0.5 \mu s/^\circ C$ , respectively, over a temperature range from 10 to  $50^\circ C$ . This means that the variation of the  $V_{G,eff}$  is about  $-2.7 \text{ mV}/^\circ C$ . The temperature dependence of the  $T_{pw}$  is significantly improved by using a specific CTAT offset voltage. A higher voltage level of  $V_{os}$  can shift the  $T_{pw}$  to a lower value.

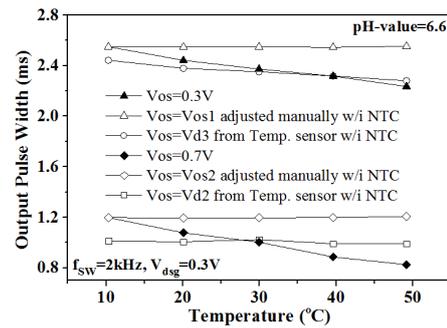


Fig. 4 Variation of the output pulse width with temperature under external constant and on-chip or external manually-setting CTAT offset voltages, respectively, for a PBS solution of pH value of 6.6.

### 3. Conclusions

A CMOS thermally stable PWM readout circuit for pH-ISFET sensors has successfully been designed and fabricated by the TSMC 0.18  $\mu m$  process. The current consumption is about 0.26 mA under the supply voltage of 1.8 V. The circuit operates under a procedure of charging of fixed time and then discharging of fixed current. Two matched V-I converters make the output characteristics of readout circuit itself thermally stable. In a solution with a pH value of 6.6, the variation of the effective input voltage,  $V_{G,eff}$ , is about  $-2.7 \text{ mV}/^\circ C$ . A suitable CTAT  $V_{os}$  can be used to effectively reduce the temperature dependence of the output pulse width related to  $H^+$  ion concentration in a solution. The sensitivity of the on-chip temperature sensor is  $-2.1 \text{ mV}/^\circ C$ . Hence the sensor provides a slightly weak temperature compensation effect. But the on-chip offset voltage still provides a good temperature compensation. The equivalent drift of pH value is less than  $0.0004/^\circ C$ .

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